



accelerating field and the synchronization of the laser. For achieving values in the 10 fs range, the field must be stabilized to 0.01% in amplitude and 0.01 deg (at 3 GHz) in phase by the LLRF system. To control the electric field of the two cavities, the real and imaginary parts of the field are measured using an intermediate frequency (IF) sampling scheme. As shown in Fig. 2, the RF signal from the cavity at 3 GHz is down-converted in an RF mixer with a local oscillator (LO) at 3.025 GHz to an IF of 25 MHz and sampled by an ADC with a sampling rate (CLK) of 125 MHz. The sampled data is preprocessed by a delay stage, IQ detection filter, and rotation matrix on the ADC board. Afterwards the data is sent to the LLRF controller. In the FPGA-based controller (Xilinx Virtex 5), the detected cavity field is compared to a given set-point. The error signal is used to generate the appropriate feed-forward signal through a model-based iterative-learning control algorithm (learning feed-forward) and a feedback signal through an intra-pulse proportional-feedback controller. The feed-forward and feedback signals are added and sent through two DACs, using an RF vector modulator and a pulse power amplifier to drive the klystron.

### The Timing System

In order to avoid 50Hz distortions from the mains in power supplies, the machine trigger is synchronized to the zero crossing of the 50 Hz. This is realized within a master timing module [4], which then delivers the trigger for the LLRF, the high power RF, the diagnostic tools, and for the amplifier system of the laser. In addition, the trigger has to be synchronous to the continuous machine frequencies from the master oscillator and the laser repetition rate. To achieve this, a base frequency ( $f_B$ ) whose zero crossing lines up with all other zero crossings is generated and the trigger is shifted in steps of periods of  $f_B$ . In this case, 1 kHz is taken as  $f_B$  because it is the lowest frequency in the system [3]. Tab. 1 is listing the values of most relevant frequencies at REGAE.

Table 1: REGAE frequencies

Main RF	$f_{RF}$	2.9979 GHz
LO for down-conversion	$f_{LO}$	3.0229 GHz
IF from down-conversion	$f_{IF}$	24.983 MHz
ADC clock	$f_s$	124.91 MHz
Reference input timing	$f_T$	0.9993 GHz
Laser osc. repetition rate	$f_{Ti:Sa}$	83.275 MHz
Laser ampl. repetition rate	$f_{las}$	1.0009 kHz
Timing base frequency	$f_B$	

### The Laser Synchronization System

A commercial Ti:Sapphire laser system is used to generate the electrons in the RF gun by frequency-tripling the

810 nm to 270 nm and to carry out the electron diffraction pump-probe experiments at the end of the beamline. The synchronization system (Fig. 3) locks the laser repetition rate to the RF master oscillator. This is accomplished by extracting the 36th harmonic of the 83.275 MHz laser pulses with a photodiode and a narrow bandpass filter, centred at 3 GHz.

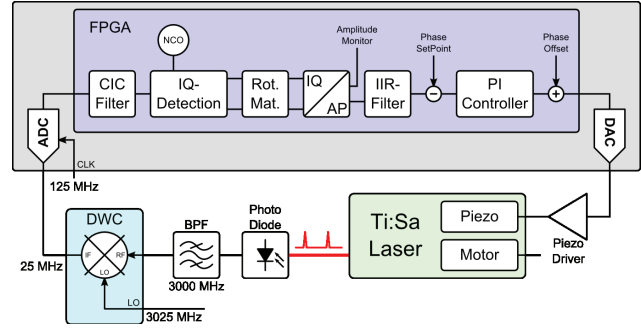


Figure 3: Layout of the digital laser synchronization system at REGAE.

The synchronization system will use the same hardware and IF sampling scheme like for the LLRF system [6]. The signal from the photodiode is down-converted, sampled and preprocessed (CIC filter, IQ detection filter, rotation matrix). The phase of the 3 GHz signal is extracted and applied in a feedback loop, which acts on a piezoelectric actuator within the laser cavity, changing frequency and phase of the optical pulse train in order to synchronize it to the master oscillator. The laser also incorporates a stepper motor for coarse tuning. In the first run of REGAE and during the measurements reported in this paper, the feedback was accomplished by a simple analog controller.

## MEASUREMENT METHOD

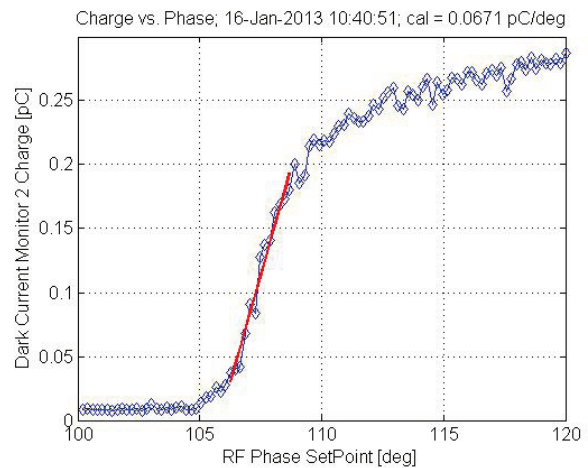


Figure 4: Charge measured versus RF gun phase during a phase scan. The linear fit (red) shows the region with the highest sensitivity (0.0671 pC/deg).

The method, which was chosen to determine the synchronization of the photo-cathode laser and the stability of the RF gun phase is described in [7]. Fig. 4 shows a so called phase scan. While the phase of the RF gun is swept, the bunch charge is measured with the dark current monitor system (DaMon) [8]. Due to its high charge resolution of below 1 fC, the DaMon at REGAE is used as a sensitive charge monitor. The red line in the plot marks the range of the RF phase, where the bunch charge is strongly dependent on the RF gun phase. With a steep linear slope of about 0.067 pC/deg it provides a sensitive measurement of the relative timing jitter between the RF gun phase and the photo-cathode laser. Obviously it can not be distinguished, if the measured jitter is induced by jitter of the RF gun phase or the timing jitter of the laser. Limitations of the measurement technique are the intensity fluctuations of the laser on the photo cathode and the resolution of the DaMon. The charge instability due to laser power fluctuations is about 1.9%, resulting in 3.2 fC charge variation at a measurement phase of 108.5 deg and 169 fC bunch charge. With 13.8 fs/fC (0.0671 pC/deg) the resolution is limited to about 44 fs rms. The resolution of the DaMon is about 0.83 fC rms, which corresponds to 11.6 fs rms.

By reference tracking (feeding the 3 GHz reference from the MO to one down-converter channel), the resolution of the RF phase measurement is determined to be below 18 fs rms at full detection bandwidth of 25 MHz. Due to the cavity bandwidth of about 400 kHz, the short pulse length of 6  $\mu$ s, and a 1  $\mu$ s latency in the control loop we are not able to get a stable intra-pulse proportional-feedback in operation. The learning feed-forward is used to reduce repetitive distortions and slow drifts in the system, which are mainly caused by temperature variation in the water system of the cavity.

## FIRST RESULTS AND FUTURE PLANS

Fig. 5 shows two measurements over a time period of 10 min. Both plots show the phase jitter of the RF gun measured by the probe antenna and derived from the charge variation at a measurement phase of 108.5 deg. In the upper graph, the learning feed-forward is off and both signals are well correlated, therefore the phase variation is strongly dominated by the RF gun. In the lower graph, the learning feed-forward is on and the phases are stabilized below 0.06 deg rms (55 fs). The residual jitter is primarily due to the measurement setup, caused by laser power fluctuation. The jitter due to the RF gun phase and the laser synchronization cannot be determined precisely. An estimate can be given by  $\sqrt{(55 \text{ fs})^2 - (44 \text{ fs})^2} = 33 \text{ fs}$  rms. Due to the missing fast feedback, all jitter contributions from the klystron chain, like additive noise from the vector modulator, pre-amplifier or klystron, are not suppressed. To reach the required synchronization of 10 fs in long and short term, further steps have to be performed. In the next half year a hardware upgrade is scheduled to the new single cavity RF control system [9]. This new hardware scheme will allow to reduce the controller latency by 500 ns and

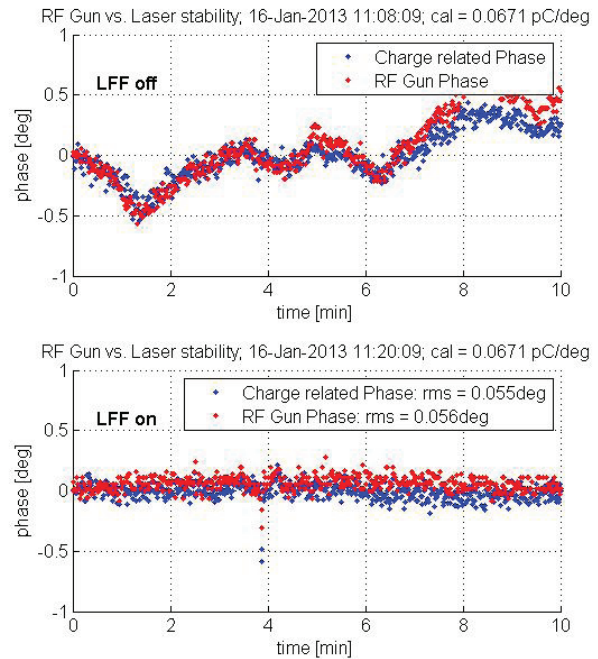


Figure 5: RF gun phase versus charge related phase with learning feed-forward turned off (top) and on (bottom).

will have higher FPGA processing power, which allows higher sampling rate and more sophisticated feedback operation to reduce actuator induced jitter. In addition, a drift calibration scheme and an RF interferometer based distribution system is planned to reduce long term phase drifts of the RF system. Furthermore a Laser-to-RF converter will be installed to perform drift free laser synchronization. In addition a photodiode is installed close to the photo cathode to remove laser induced charge fluctuations from the measurements.

## REFERENCES

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